

A Strong Converse for Joint Source-Channel Coding

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Abstract—We consider a discrete memoryless joint source-channel setting. Let D be some distortion level, lower than the optimum performance theoretically attainable according to the separation theorem. We prove that for any joint source-channel scheme, the probability that the distortion is at most D approaches zero as the block length increases. Furthermore, we show that the probability has an exponential behavior, and evaluate the optimal exponent.

I. INTRODUCTION

Information Theory produces sharp results: if the required performance is below a threshold, set by the problem parameters, then it may be achieved reliably, i.e., with probability that approaches one; otherwise, it can not. Converse results may be divided into categories, according to their strength, measured asymptotically as the block length goes to infinity. A weak converse indicates it is impossible for the probability of success to approach one. A strong converse further states that the success probability must approach zero. Beyond that, one may be interested in the rate in which this happens; specifically, an “exponentially-strong” converse states that the probability of success must approach zero exponentially. Indeed, for source and channel coding exponentially strong converses are known.

For a discrete memoryless source (DMC) W , when the required channel rate R is above the channel capacity C , the probability of correct decoding decays with the following exponent [1] (Earlier in [2] it appears in a different form and proven to be a lower bound on the exponent):¹

$$\bar{E}_{\text{sp}}(R, W) = \max_{\Phi} \min_V \left[D(V \| W | \Phi) + |R - I(\Phi, V)|^+ \right]. \quad (1)$$

This exponent is closely related the sphere-packing exponent, which is an upper bound on the exponent of the error probability when $R < C$ [3] (in this form, [4]):

$$E_{\text{sp}}(R, W) \triangleq \max_{\Phi} \min_{V: I(\Phi, V) \leq R} D(V \| W | \Phi).$$

In lossy source coding, dual results can be obtained where the *excess-distortion probability* plays the role of error probability. Specifically, for some single-letter distortion measure $d(\cdot, \cdot)$ and distortion threshold D , we say that the scheme was successful if

$$\frac{1}{n} \sum_{i=1}^n d(S, \hat{S}) \leq D.$$

¹For definition of divergence and mutual information, see notation section in the sequel

For a discrete memoryless source (DMS) P , when the rate R is below the rate-distortion function (RDF) $R(P, D)$, the success probability has exponent [5, Problem 9.6]:

$$\bar{E}_{\text{S}}(R, D, P) = \min_Q \left[D(Q \| P) + |R(Q, D) - R|^+ \right], \quad (2)$$

Similar to the channel coding case, this exponent is closely related to the lossy source coding excess-distortion exponent for $R > R(P, D)$ [6]:

$$E_{\text{S}}(R, D, P) = \inf_{Q: R(Q, D) > R} D(Q \| P).$$

In joint source-channel coding (JSCC), an average distortion D is achievable if $R(D) < \rho C$ and not achievable if the opposite holds, where ρ is the *bandwidth expansion factor* (number of channel uses per source sample). This result, due to Shannon [7] immediately implies a weak converse for the excess-distortion probability. However, to the best of our knowledge, a strong converse has not been presented, except for [8], where Zhong et al. use very specific arguments to prove it for the Quadratic-Gaussian setting. Indeed, such a strong converse may be derived using previously known results. One way to do so, is to use equivalence to channel coding [9] and the strong channel converse. Alternatively, JSCC dispersion [10] implies a strong converse. Information spectrum methods, such as used by Han [11] to derive a strong converse to lossless JSCC.

In this work we take a direct path that allows us not only to prove a strong converse, but also derive the exponential behavior of the probability of success. Specifically, we show that whenever $R(D) > \rho C$, the probability of not having excess distortion for the optimal JSCC scheme decays with exponent

$$\begin{aligned} & \bar{E}_{\text{JSCC}}(P, D, W, \rho) \\ &= \min_R [\bar{E}_{\text{S}}(R, D, P) + \rho \bar{E}_{\text{sp}}(R/\rho, W)], \end{aligned} \quad (3)$$

where the channel and source exponents are given by (1) and (2), respectively. This is analogous to the exponent of the error probability: the JSCC excess-distortion exponent for $R(D) < \rho C$ is upper-bounded by [12], [13]

$$\min_R [E_{\text{S}}(R, D, P) + \rho E_{\text{sp}}(R, W)]. \quad (4)$$

II. NOTATIONS

This paper uses lower case letters (e.g. x) to denote a particular value of the corresponding random variable denoted in capital letters (e.g. X). Vectors are denoted in bold (e.g. \mathbf{x} or \mathbf{X}). Calligraphic fonts (e.g. \mathcal{X}) represent a set and $\mathcal{P}(\mathcal{X})$ denotes all the probability distributions on the alphabet \mathcal{X} . We use \mathbb{Z}_+ and \mathbb{R}_+ to denote the set of non-negative integer and real numbers respectively.

Our proofs make use of the method of types, and follow the notations in [5]. Specifically, the *type* of a sequence \mathbf{x} with length n is denoted by $P_{\mathbf{x}}$, where the type is the empirical distribution of this sequence, i.e., $P_{\mathbf{x}}(a) = N(a|\mathbf{x})/n \forall a \in \mathcal{X}$, where $N(a|\mathbf{x})$ is the number of occurrences of a in sequence \mathbf{x} . The subset of the probability distributions $\mathcal{P}(\mathcal{X})$ that can be types of n -sequences is denoted as

$$\mathcal{P}_n(\mathcal{X}) \triangleq \{P \in \mathcal{P}(\mathcal{X}) : nP(x) \in \mathbb{Z}_+, \forall x \in \mathcal{X}\} \quad (5)$$

and sometimes P_n is used to emphasize the fact that $P_n \in \mathcal{P}_n(\mathcal{X})$. A *type class* $\mathcal{T}_{P_{\mathbf{x}}}^n$ is defined as the set of sequences that have type $P_{\mathbf{x}}$. Given some sequence \mathbf{x} , a sequence \mathbf{y} of the same length has *conditional type* $P_{\mathbf{y}|\mathbf{x}}$ if $N(a, b|\mathbf{x}, \mathbf{y}) = P_{\mathbf{y}|\mathbf{x}}(a|b)N(a|\mathbf{x})$. Furthermore, the random variable corresponding to the conditional type of a random vector \mathbf{Y} given \mathbf{x} is denoted as $P_{\mathbf{Y}|\mathbf{x}}$. In addition, the possible conditional type given an input distribution $P_{\mathbf{x}}$ is denoted as

$$\mathcal{P}_n(\mathcal{Y}|P_{\mathbf{x}}) \triangleq \{P_{\mathbf{y}|\mathbf{x}} : P_{\mathbf{x}} \times P_{\mathbf{y}|\mathbf{x}} \in \mathcal{P}_n(\mathcal{X} \times \mathcal{Y})\}.$$

A discrete memoryless channel (DMC) $W : \mathcal{X} \rightarrow \mathcal{Y}$ is defined with its input alphabet \mathcal{X} , output alphabet \mathcal{Y} , and conditional distribution $W(\cdot|x)$ of output letter Y when the channel input letter X equals $x \in \mathcal{X}$. Also, we abbreviate $W(\cdot|x)$ as $W_x(\cdot)$ for notational simplicity. We define mutual information as

$$I(\Phi, W) = \sum_{x,y} \Phi(x)W(y|x) \log \frac{\Phi(x)W(y|x)}{\Phi W(y)},$$

and the channel capacity is given by

$$C(W) = \max_{\Phi} I(\Phi, W),$$

and the set of capacity-achieving distributions is $\Pi(W) \triangleq \{\Phi : I(\Phi, W) = C(W)\}$.

A discrete memoryless source (DMS) is defined with source alphabet \mathcal{S} , reproduction alphabet $\hat{\mathcal{S}}$, source distribution P and a distortion measure $d : \mathcal{S} \times \hat{\mathcal{S}} \rightarrow \mathbb{R}_+$. Without loss of generality, we assume that for any $s \in \mathcal{S}$ there is $\hat{s} \in \hat{\mathcal{S}}$ such that $d(s, \hat{s}) = 0$. The rate-distortion function (RDF) of a DMS $(\mathcal{S}, \hat{\mathcal{S}}, P, d)$ is given by

$$R(P, D) = \min_{\Lambda : E_{P, \Lambda} d(S, \hat{S}) \leq D} I(P, \Lambda),$$

where

$$d(\mathbf{s}, \hat{\mathbf{s}}) \triangleq \frac{1}{n} \sum_{i=1}^n d(s_i, \hat{s}_i) \quad (6)$$

is the distortion between the source and reproduction words \mathbf{s} and $\hat{\mathbf{s}}$ and $I(P, \Lambda)$ is the mutual information over a channel with input distribution $P(S)$ and conditional distribution $\Lambda : \mathcal{S} \rightarrow \hat{\mathcal{S}}$.

A discrete memoryless joint source-channel coding (JSCC) problem consists of a DMS $(\mathcal{S}, \hat{\mathcal{S}}, P, d)$, a DMC $W : \mathcal{X} \rightarrow \mathcal{Y}$ and a *bandwidth expansion factor* $\rho \in \mathbb{R}_+$. A JSCC scheme $\mathcal{C}_{\text{JSCC}}^{(n)}$ is comprised of an encoder mapping $f_{J;n} : \mathcal{S}^n \rightarrow \mathcal{X}^{\lfloor \rho n \rfloor}$ and decoder mapping $g_{J;n} : \mathcal{Y}^{\lfloor \rho n \rfloor} \rightarrow \hat{\mathcal{S}}^n$. Given a source block \mathbf{s} , the encoder maps it to a sequence $\mathbf{x} = f_{J;n}(\mathbf{s}) \in \mathcal{X}^{\lfloor \rho n \rfloor}$ and transmits this sequence through the channel. The decoder receives a sequence $\mathbf{y} \in \mathcal{Y}^{\lfloor \rho n \rfloor}$ distributed according to $W(\cdot|\mathbf{x})$, and maps it to a source reconstruction $\hat{\mathbf{s}}$. The corresponding distortion is given by (6). For a given JSCC scheme, we define the *error event* $\mathcal{E}(D)$ as

$$\mathcal{E}(D) \triangleq \mathcal{E}(D, f_{J;n}, g_{J;n}) \quad (7)$$

$$\triangleq \mathcal{E}(D, \mathcal{C}_{\text{JSCC}}^{(n)}) \quad (8)$$

$$\triangleq \{d(\mathbf{S}, \hat{\mathbf{S}}) > D\}, \quad (9)$$

and the *correct event* $\bar{\mathcal{E}}(D) \triangleq \mathcal{E}(D)^c = \{d(\mathbf{S}, \hat{\mathbf{S}}) \leq D\}$. Finally, for block length n , we define the *best correct event* $\bar{\mathcal{E}}_n(D)$ as an event that corresponds to the JSCC scheme that produces the minimum error probability, i.e.,

$$\bar{\mathcal{E}}_n(D) \in \arg \min_{\{\mathcal{C}_{\text{JSCC}}^{(n)}\}} \mathbb{P}[\bar{\mathcal{E}}(D)].$$

III. MAIN RESULT

The following formally states the exponential decay rate of the probability of success at distortion thresholds $R(D) > \rho C$, thus also serves as a strong converse for JSCC coding.

Theorem 1 (Strong Converse for JSCC). *Given a discrete memoryless JSCC problem with DMS $(\mathcal{S}, \hat{\mathcal{S}}, P, d)$ DMC $(\mathcal{X}, \mathcal{Y}, W)$ and bandwidth expansion factor ρ , when $\rho C(W) < R(P, D)$, let $\bar{E}(P, D, W, \rho)$ be the exponent of the success probability for the best sequence of JSCC schemes*

$$\bar{E}(P, D, W, \rho) \triangleq \lim_{n \rightarrow \infty} -\frac{1}{n} \log \mathbb{P}[\bar{\mathcal{E}}_n(D)].$$

Then $\bar{E}(P, D, W, \rho)$ is given by $\bar{E}_{\text{JSCC}}(P, D, W, \rho)$ (3).

Remark 1 (Direct part of the theorem). *The achievability of the exponent $\bar{E}_{\text{JSCC}}(P, D, W, \rho)$ may be proven by using unequal error protection (UEP), as done by Csiszár for the JSCC exponent where $R(D) < \rho C$ [13]. Loosely speaking, each source type-class Q is quantized by a codebook of rate $R(Q, D)$, and then a UEP scheme is used to transmit these codebooks over the channel. The achievability proof [1] of the channel exponent $\bar{E}_{\text{sp}}(R, W)$ in (1) can be extended to the UEP setting, completing the proof. As this is a rather trivial extension of previous results, we concentrate in this work on proving the converse.*

Remark 2 (Alternative form). *The exponent $\bar{E}_{\text{JSCC}}(P, D, W, \rho)$ may be written explicitly as a function of*

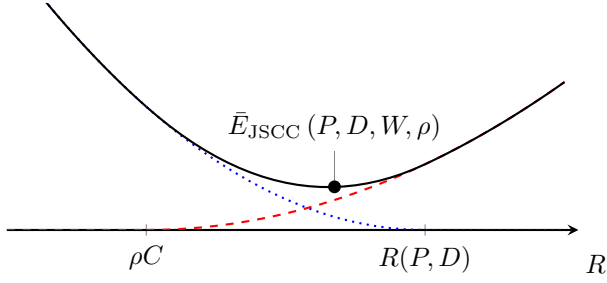


Fig. 1. When $R(P, D) > \rho C$, (3) is always minimized by a rate R such that $R(P, D) > R > \rho C$, where in the plot the dashed curve is $\bar{E}_{\text{sp}}(R, W)$, the dotted curve is $\bar{E}_{\text{S}}(R, D, P)$ and the solid curve is $\bar{E}_{\text{S}}(R, D, P) + \rho \bar{E}_{\text{sp}}(R, W)$.

the source and channel parameters, as follows.

$$\begin{aligned} & \bar{E}_{\text{JSCC}}(P, D, W, \rho) \\ &= \min_{Q \in \mathcal{P}(S)} \left[D(Q \| P) + \max_{\Phi \in \mathcal{P}(\mathcal{X})} \min_{V \in \mathcal{P}(\mathcal{Y}|\mathcal{X})} \left(\rho D(V \| W|\Phi) \right. \right. \\ & \quad \left. \left. + |R(Q, D) - \rho I(\Phi, V)|^+ \right) \right]. \end{aligned} \quad (10)$$

We prove the equivalence of this form to (3) in Section IV-C. We use this form in the proof of Theorem 1, rather than the form (3).

Remark 3 (Minimizing rate). When $R(P, D) > \rho C$, the rate minimizing (3) satisfies $R(P, D) \geq R \geq \rho C$, as shown in Fig. 1. This is parallel to the excess-distortion exponent where $R(P, D) < \rho C$, where (4) is minimized by a rate $R(P, D) < R < \rho C$.

The proof of our main result builds on the following key lemma.

Lemma 2 (Joint source channel coding converse with fixed types). For a JSCC problem, given a source type $Q \in \mathcal{P}_n(S)$ and a channel input type $\Phi \in \mathcal{P}_n(\mathcal{X})$, define all the channel outputs that covers \mathbf{s} with distortion D as $\hat{B}(\mathbf{s}, D)$, i.e.,

$$\hat{B}(\mathbf{s}, D) \triangleq \{\mathbf{y} \in \mathcal{Y}^m : d(\mathbf{s}, g_{J;n}(\mathbf{y})) \leq D\} \quad (11)$$

where $m = \lfloor \rho n \rfloor$ and $g_{J;n}$ is the JSCC decoder. Then for a given distortion D and a channel with constant composition conditional distribution $V \in \mathcal{P}_m(\mathcal{Y}|\Phi)$, we have

$$\begin{aligned} & \frac{1}{|G(Q, \Phi)|} \sum_{\mathbf{s}_i \in G(Q, \Phi)} \frac{|\mathcal{T}_V^m(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D)|}{|\mathcal{T}_V^m(f(\mathbf{s}_i))|} \\ & \leq \frac{p(n)}{\alpha(Q, \Phi)} \exp^{-n[R(Q, D) - \rho I(\Phi, V)]^+}, \end{aligned} \quad (12)$$

where $p(n)$ is a polynomial that depends only on the source, channel and reconstruction alphabet sizes and ρ .

The proof of this lemma is based upon the exponential channel coding converse [1], combined with the following.

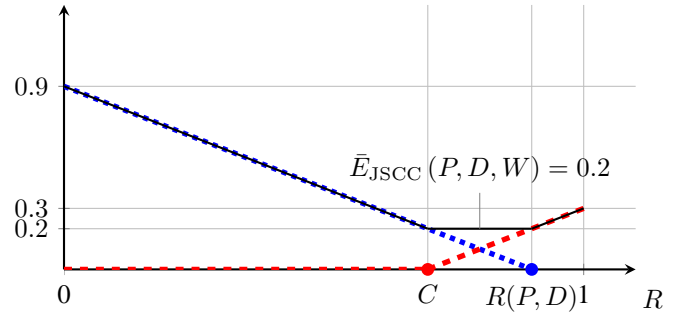


Fig. 2. $\bar{E}_{\text{JSCC}}(P, D, W, \rho)$ with $H_b(D) = 0.1$ and $H_b(\varepsilon) = 0.3$, where in the plot the dashed curve is $\bar{E}_{\text{sp}}(R, W)$, the dotted curve is $\bar{E}_{\text{S}}(R, D, P)$ and the solid curve is $\bar{E}_{\text{S}}(R, D, P) + \bar{E}_{\text{sp}}(R, W)$.

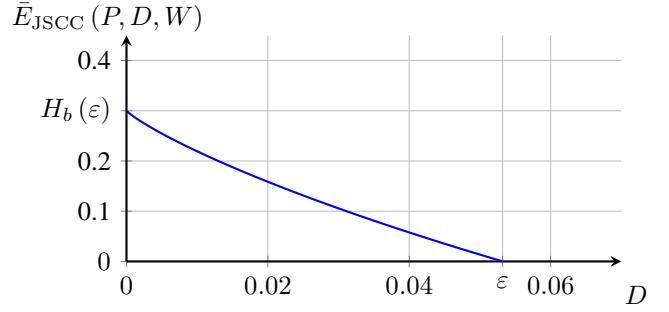


Fig. 3. $\bar{E}(P, D, W, \rho)$ with $H_b(\varepsilon) = 0.3$.

Lemma 3 (Restricted D -ball size). Given source type P and a reconstruction sequence $\hat{\mathbf{s}}$, define restricted D -ball as

$$B(\hat{\mathbf{s}}, P, D) \triangleq \{\mathbf{s} \in \mathcal{T}_P^n : d(\mathbf{s}, \hat{\mathbf{s}}) \leq D\}.$$

Then

$$|B(\hat{\mathbf{s}}, P, D)| \leq (n+1)^{|\mathcal{S}|} \exp\{n[H(P) - R(P, D)]\}$$

This last lemma is similar to Lemma 3 in [14]. However, it provides a *uniform* bound over the size of the restricted D -ball. We prove this lemma, as well as Lemma 2 and the converse part of Theorem 1 in the next section. We conclude this section by presenting the following example.

Example 1. Transmitting a binary symmetric source (BSS) over a binary symmetric channel (BSC) subject to the Hamming distortion with the bandwidth expansion factor $\rho = 1$.

For a BSS, the RDF is given by $R(P, D) = 1 - H_b(D)$, where $H_b(\cdot)$ is the binary entropy function. It can be shown that (1) is always minimized by a uniform distribution and:

$$\bar{E}_{\text{S}}(R, D, P) = |1 - H_b(D) - R|^+.$$

For a BSC with cross over probability ε , the capacity is given by $C(W) = 1 - H_b(\varepsilon)$. It can be shown that the optimizing Φ and V in (1) are always symmetric and:

$$\bar{E}_{\text{sp}}(R, W) = |R - 1 + H_b(\varepsilon)|^+.$$

Therefore, when $R(P, D) > C(W)$, i.e. $D < \varepsilon$,

$$\begin{aligned}\bar{E}_{\text{JSCC}}(P, D, W) &= \inf_R [\bar{E}_S(R, D, P) + \bar{E}_{\text{sp}}(R, W)] \\ &= R(P, D) - C(W) \\ &= H_b(\varepsilon) - H_b(D).\end{aligned}$$

For the case of $H_b(D) = 0.1$ and $H_b(\varepsilon) = 0.3$, we plot the $\bar{E}_S(R, D, P)$, $\bar{E}_{\text{sp}}(R, W)$ and $\bar{E}_{\text{JSCC}}(P, D, W)$ in Fig. 2. Finally, we show how $\bar{E}_{\text{JSCC}}(P, D, W)$ as a function of D when $H_b(\varepsilon) = 0.3$ in Fig. 3.

IV. PROOFS

A. Proof of Key Lemmas

Proof of Lemma 3: Let $P \in \mathcal{P}_n(\mathcal{S})$ be a given type and let Q be the type of $\hat{\mathbf{s}}$. Then the size of the set of source codewords with type P that are D -covered by $\hat{\mathbf{s}}$ is

$$|B(\hat{\mathbf{s}}, P, D)| = \left| \bigcup_{\substack{\Lambda: \mathbb{E}[d(\mathcal{S}, \hat{\mathcal{S}})] \leq D, \\ P\Lambda = Q}} \{\mathbf{s} \in \mathcal{T}_P^n : P_{\mathbf{s}, \hat{\mathbf{s}}} = P \times \Lambda\} \right|.$$

Note there are at most $(n+1)^{|\mathcal{S}||\hat{\mathcal{S}}|}$ joint types, and

$$\{\mathbf{s} \in \mathcal{T}_P^n : P_{\mathbf{s}, \hat{\mathbf{s}}} = P \times \Lambda\} = \mathcal{T}_{\tilde{\Lambda}}^n(\hat{\mathbf{s}}),$$

where $\tilde{\Lambda}$ is the reverse channel from $\hat{\mathcal{S}}$ to \mathcal{S} such that $Q \times \tilde{\Lambda} = P \times \Lambda$. Therefore,

$$\begin{aligned}|B(\hat{\mathbf{s}}, P, D)| &\leq \sum_{\tilde{\Lambda}: \mathbb{E}_{Q, \tilde{\Lambda}}[d(\hat{\mathcal{S}}, \mathcal{S})] \leq D} |\mathcal{T}_{\tilde{\Lambda}}^n(\hat{\mathbf{s}})| \\ &\leq (n+1)^{|\mathcal{S}||\hat{\mathcal{S}}|} \exp \left[n \max_{\tilde{\Lambda}: \mathbb{E}_{Q, \tilde{\Lambda}}[d(\hat{\mathcal{S}}, \mathcal{S})] \leq D} H(\tilde{\Lambda}|Q) \right].\end{aligned}$$

Note

$$\begin{aligned}R(P, D) &= \min_{\Lambda: \mathbb{E}_{P, \Lambda}[d(\mathcal{S}, \hat{\mathcal{S}})] \leq D} I(P, \Lambda) \\ &= H(P) - \max_{\tilde{\Lambda}: \mathbb{E}_{Q, \tilde{\Lambda}}[d(\hat{\mathcal{S}}, \mathcal{S})] \leq D} H(\tilde{\Lambda}|Q),\end{aligned}$$

hence

$$|B(\hat{\mathbf{s}}, P, D)| \leq (n+1)^{|\mathcal{S}||\hat{\mathcal{S}}|} \exp \{n[H(P) - R(P, D)]\}.$$

Proof for Lemma 2: In our proof, we first bound the denominator in (12) uniformly for all \mathbf{s}_i , and then bound the sum of the numerator over all \mathbf{s}_i , as done in [1] for the channel error exponent.

a) *Bounding the denominator:* Based on standard results in method of types [5], for $f(\mathbf{s}) \in \mathcal{T}_{\Phi}^n$,

$$(m+1)^{-|\mathcal{X}||\mathcal{Y}|} \exp \{mH(V|\Phi)\} \leq |\mathcal{T}_V^m(f(\mathbf{s}))|.$$

Hence

$$\frac{1}{|\mathcal{T}_V^m(f(\mathbf{s}))|} \leq (m+1)^{|\mathcal{X}||\mathcal{Y}|} \exp \{-mH(V|\Phi)\}.$$

b) *Bounding the sum of numerator:* Since $\mathbf{s} \in G(Q, \Phi)$,

$$\begin{aligned}\mathbf{y} &\in \mathcal{T}_V(f(\mathbf{s})) \cap \hat{B}(\mathbf{s}, D) \\ \Rightarrow \mathbf{s} &\in B(g_{J;n}(\mathbf{y}), Q, D) \cap G(Q, \Phi).\end{aligned}\quad (13)$$

Therefore, any \mathbf{y} will be counted at most $|B(g_{J;n}(\mathbf{y}), Q, D) \cap G(Q, \Phi)|$ times. According to Lemma 3, this is upper bounded by $B_u = (n+1)^{|\mathcal{S}||\hat{\mathcal{S}}|} \exp \{n[H(Q) - R(Q, D)]\}$. In addition, it is obvious that

$$\bigcup_{\mathbf{s}_i \in G(Q, \Phi)} \mathcal{T}_V(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D) \subset \mathcal{T}_{\Psi}^n,$$

where $\Psi = \Phi V$ is the channel output distribution corresponding to Φ . Therefore,

$$\begin{aligned}&\frac{\alpha(Q, \Phi)}{|G(Q, \Phi)|} \sum_{\mathbf{s}_i \in G(Q, \Phi)} |\mathcal{T}_V(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D)| \\ &\leq \frac{1}{|\mathcal{T}_Q^n|} B_u \left| \bigcup_{\mathbf{s}_i \in \mathcal{T}_Q^n} \mathcal{T}_V(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D) \right| \\ &\leq \frac{1}{|\mathcal{T}_Q^n|} B_u |\mathcal{T}_{\Psi}^n|.\end{aligned}$$

Noting

$$\begin{aligned}(n+1)^{-|\mathcal{S}|} \exp \{nH(Q)\} &\leq |\mathcal{T}_Q^n| \\ |\mathcal{T}_{\Psi}^m| &\leq \exp \{mH(\Psi)\},\end{aligned}$$

we have

$$\begin{aligned}&\frac{1}{n} \log \left[\frac{1}{|\mathcal{T}_Q^n|} B_u |\mathcal{T}_{\Psi}^n| \right] \\ &\leq \frac{|\mathcal{S}|}{n} \log(n+1) - H(Q) \\ &\quad + \frac{|\mathcal{S}||\hat{\mathcal{S}}|}{n} \log(n+1) + H(Q) - R(Q, D) \\ &\quad + \rho H(\Psi) \\ &\leq \rho H(\Psi) - R(Q, D) \\ &\quad + \frac{|\mathcal{S}||\hat{\mathcal{S}}|}{n} \log(n+1) + \frac{|\mathcal{S}|}{n} \log(n+1).\end{aligned}$$

Combining the bounds for both numerator and denominator, we have

$$\begin{aligned}&\frac{1}{n} \log \left[\frac{\alpha(Q, \Phi)}{|G(Q, \Phi)|} \sum_{\mathbf{s}_i \in G(Q, \Phi)} \frac{|\mathcal{T}_V(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D)|}{|\mathcal{T}_V(f(\mathbf{s}_i))|} \right] \\ &\leq \rho H(\Psi) - \rho H(V|\Phi) - R(Q, D) + \frac{|\mathcal{X}||\mathcal{Y}|}{m} \log(m+1) \\ &\quad + \frac{|\mathcal{S}||\hat{\mathcal{S}}|}{n} \log(n+1) + \frac{|\mathcal{S}|}{n} \log(n+1).\end{aligned}$$

Note $m = \lfloor \rho n \rfloor \leq \rho n$, let

$$p(n) = (\rho n + 1)^{\rho|\mathcal{X}||\mathcal{Y}|} (n+1)^{(|\mathcal{S}||\hat{\mathcal{S}}|+|\mathcal{S}|)}, \quad (14)$$

and the proof is completed. \blacksquare

B. Proof of Main Result

Proof for Theorem 1 (converse part): Let $\mathbb{P}[\bar{\mathcal{E}}(D)] = 1 - \mathbb{P}[\mathcal{E}(D)]$. By following a similar argument in [1, Proof of Lemma 5], clearly,

$$\mathbb{P}[\bar{\mathcal{E}}(D)] \leq (n+1)^{|\mathcal{S}|} \max_{Q \in \mathcal{P}_n(\mathcal{S})} \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q] e^{-nD(Q \| P)}. \quad (15)$$

Let $m \triangleq \lfloor \rho m \rfloor$ and let \mathcal{A}_m denote the channel codebook in this JSCC scheme. Then conditioning again, we have

$$\begin{aligned} & \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q] \\ &= \sum_{\Phi \in \mathcal{A}_m} \mathbb{P}[P_{\mathbf{X}} = \Phi | P_{\mathbf{S}} = Q] \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q, P_{\mathbf{X}} = \Phi]. \end{aligned}$$

Let $G(Q, \Phi)$ be the set of source sequences in \mathcal{T}_Q^n that are mapped (via JSCC encoder $f_{J;n}$) to channel codewords with type Φ , i.e.,

$$G(Q, \Phi) \triangleq \{\mathbf{s} \in \mathcal{T}_Q^n : \mathbf{x} = f_{J;n}(\mathbf{s}) \in \mathcal{T}_{\Phi}^m\},$$

and let $\alpha(Q, \Phi) \triangleq \mathbb{P}[P_{\mathbf{X}} = \Phi | P_{\mathbf{S}} = Q]$ be the probability of having channel input type Φ given that the source type is Q . Noting that given a source type, all strings within a type class are equally likely, hence

$$\alpha(Q, \Phi) = \frac{|\{\mathbf{s} \in \mathcal{T}_Q^n : \mathbf{x} = f_{J;n}(\mathbf{s}) \in \mathcal{T}_{\Phi}^m\}|}{|\mathcal{T}_Q^n|} = \frac{|G(Q, \Phi)|}{|\mathcal{T}_Q^n|}.$$

Therefore,

$$\begin{aligned} & \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q] \\ &= \sum_{\Phi \in \mathcal{A}_m} \alpha(Q, \Phi) \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q, P_{\mathbf{X}} = \Phi] \\ &= \sum_{\Phi \in \mathcal{A}_m} \alpha(Q, \Phi) \left(\sum_{V \in \mathcal{P}_m(\mathcal{Y}|\Phi)} \mathbb{P}[P_{\mathbf{Y}|\mathbf{X}} = V | P_{\mathbf{X}} = \Phi] \right. \\ & \quad \left. \cdot \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q, P_{\mathbf{X}} = \Phi, P_{\mathbf{Y}|\mathbf{X}} = V] \right). \end{aligned}$$

Next we use Lemma 2 to assert,

$$\begin{aligned} & \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q, P_{\mathbf{X}} = \Phi, P_{\mathbf{Y}|\mathbf{X}} = V] \\ & \leq \frac{1}{|G(Q, \Phi)|} \sum_{\mathbf{s}_i \in G(Q, \Phi)} \frac{|\mathcal{T}_V^m(f(\mathbf{s}_i)) \cap \hat{B}(\mathbf{s}_i, D)|}{|\mathcal{T}_V^m(f(\mathbf{s}_i))|} \\ & \leq \frac{p(n)}{\alpha(Q, \Phi)} \exp\left\{-n |R(Q, D) - \rho I(\Phi, V)|^+\right\} \end{aligned}$$

where $p(n)$ is a polynomial given in (14). Therefore,

$$\begin{aligned} & \mathbb{P}[\bar{\mathcal{E}}(D) | P_{\mathbf{S}} = Q] \\ & \leq \sum_{\Phi \in \mathcal{A}_m} \alpha(Q, \Phi) \sum_{V \in \mathcal{P}_m(\mathcal{Y}|\Phi)} \mathbb{P}[P_{\mathbf{Y}|\mathbf{X}} = V | P_{\mathbf{X}} = \Phi] \left(\frac{p(n)}{\alpha(Q, \Phi)} \right. \\ & \quad \left. \cdot \exp\{-n |R(Q, D) - \rho I(\Phi, V)|\} \right) \\ & \leq \sum_{\Phi \in \mathcal{A}_m} p(n) \exp\left\{-m \left[\min_{V \in \mathcal{P}_m(\mathcal{Y}|\Phi)} D(V \| W|\Phi) \right] \right\} \\ & \quad \cdot \exp\left\{-n |R(Q, D) - \rho I(\Phi, V)|^+\right\} \\ & \leq \sum_{\Phi \in \mathcal{A}_m} p(n) \cdot \exp\left\{-n \left[\min_{V \in \mathcal{P}_m(\mathcal{Y}|\Phi)} \rho D(V \| W|\Phi) \right. \right. \\ & \quad \left. \left. + |R(Q, D) - \rho I(\Phi, V)|^+ \right] \right\} \quad (16) \end{aligned}$$

Combining (15) and (16), we have

$$\begin{aligned} & \mathbb{P}[\bar{\mathcal{E}}(D)] \\ & \leq \text{poly}(n) \exp\left\{-n \left[\min_{Q \in \mathcal{P}_n(\mathcal{S})} \left(D(Q \| P) \right. \right. \right. \\ & \quad \left. \left. + \min_{\Phi \in \mathcal{A}_m} \min_{V \in \mathcal{P}_m(\mathcal{Y}|\Phi)} \rho D(V \| W|\Phi) \right. \right. \\ & \quad \left. \left. + |R(Q, D) - \rho I(\Phi, V)|^+ \right) \right] \right\}. \end{aligned}$$

Therefore,

$$\begin{aligned} & \bar{E}(P, D, W, \rho) \\ &= \liminf_{n \rightarrow \infty} -\frac{1}{n} \log \mathbb{P}[\bar{\mathcal{E}}_n(D)] \\ & \geq \min_{Q \in \mathcal{P}(\mathcal{S})} \left(D(Q \| P) + \max_{\Phi \in \mathcal{P}(\mathcal{X})} \min_{V \in \mathcal{P}_n(\mathcal{Y}|\Phi)} \rho D(V \| W|\Phi) \right. \\ & \quad \left. + |R(Q, D) - \rho I(\Phi, V)|^+ \right). \end{aligned}$$

On account of the exponent equivalence, the proof is completed. \blacksquare

C. Proof of Exponent Equivalence

In this section we prove the equivalence between (10) and (3).

To that end, let the Φ, V and Q that achieves $\bar{E}_{\text{JSCC}}(P, D, W, \rho)$ be Φ^*, V^* and Q^* . Let R_{S} and R_{C} be the source and channel rate respectively, where $R_{\text{C}} = R_{\text{S}}/\rho$. If $R(Q^*, D) - \rho I(\Phi^*, V^*) \leq 0$, then

$$\begin{aligned} & \bar{E}_{\text{JSCC}}(P, D, W, \rho) \\ &= D(Q^* \| P) + \rho D(V^* \| W|\Phi^*) \\ &= \min_Q D(Q \| P) + \max_{\Phi} \min_V \rho D(V \| W|\Phi), \end{aligned}$$

and for any $R(Q^*, D) \leq R'_{\text{S}} \leq \rho I(\Phi^*, V^*)$ and $R'_{\text{C}} = R'_{\text{S}}/\rho$, we have

$$\begin{aligned} & \bar{E}_{\text{S}}(R'_{\text{S}}, D, P) = \min_Q [D(Q \| P)], \\ & \bar{E}_{\text{sp}}(R'_{\text{C}}, W) = \max_{\Phi} \min_V D(V \| W|\Phi), \end{aligned}$$

and thus

$$\begin{aligned} & \inf_{R_S} \bar{E}_S(R_S, D, P) + \rho \bar{E}_{\text{sp}}(R_C, W) \\ & \leq \bar{E}_S(R'_S, D, P) + \rho \bar{E}_{\text{sp}}(R'_C, W) \\ & = \bar{E}_{\text{JSCC}}(P, D, W, \rho). \end{aligned} \quad (17)$$

Similarly, we can show that when $R(Q^*, D) - \rho I(\Phi^*, V^*) > 0$,

$$\begin{aligned} & \inf_{R_S} \bar{E}_S(R_S, D, P) + \rho \bar{E}_{\text{sp}}(R_C, W) \\ & \leq \bar{E}_{\text{JSCC}}(P, D, W, \rho). \end{aligned} \quad (18)$$

However, we also know that for any R

$$\begin{aligned} & \bar{E}_S(R_S, D, P) + \rho \bar{E}_{\text{sp}}(R_C, W) \\ & \geq \min_{Q \in \mathcal{P}(\mathcal{S})} \left(D(Q \| P) + \max_{\Phi \in \mathcal{P}(\mathcal{X})} \min_{V \in \mathcal{P}(\mathcal{Y}|\mathcal{X})} \rho D(V \| W | \Phi) \right. \\ & \quad \left. + |R(Q, D) - \rho I(\Phi, V)|^+ \right), \end{aligned}$$

hence

$$\inf_{R_S} \bar{E}_S(R_S, D, P) + \rho \bar{E}_{\text{sp}}(R_C, W) \geq \bar{E}_{\text{JSCC}}(P, D, W, \rho). \quad (19)$$

Therefore, (17) and (18) and (19) suggest that

$$\bar{E}_{\text{JSCC}}(P, D, W, \rho) = \inf_{R_S} \bar{E}_S(R_S, D, P) + \rho \bar{E}_{\text{sp}}(R_C, W).$$

REFERENCES

- [1] G. Dueck and J. Korner. Reliability function of a discrete memoryless channel at rates above capacity. *IEEE Trans. Info. Theory*, 25(1):82–85, January 1979.
- [2] S. Arimoto. On the converse to the coding theorem for discrete memoryless channels (corresp.). *IEEE Trans. Info. Theory*, 19(3):357–359, may 1973.
- [3] C. E. Shannon, R. G. Gallager, and E. R. Berlekamp. Lower bounds to error probability for coding on discrete memoryless channels, part I-II. *Inform. Contr.*, 10:65–103,522–552, 1967.
- [4] E. A. Haroutunian. Estimates of the error exponent for the semi-continuous memoryless channel. *Problemy Pered. Inform. (Problems of Inform. Trans.)*, 4, No. 4:37–48, 1968.
- [5] I. Csiszár and J. Korner. *Information Theory - Coding Theorems for Discrete Memoryless Systems*. Cambridge University Press, New York, 2011.
- [6] K. Marton. Error exponent for source coding with a fidelity criterion. *IEEE Trans. Info. Theory*, IT-20:197–199, Mar. 1974.
- [7] C. E. Shannon. Coding theorems for a discrete source with a fidelity criterion. In *Institute of Radio Engineers, International Convention Record, Vol. 7*, pages 142–163, 1959.
- [8] Y. Zhong, F. Alajaji, and L.L. Campbell. On the excess distortion exponent for memoryless Gaussian source-channel pairs. In *ISIT-2006, Seattle, WA, 2006*.
- [9] M. Agarwal, A. Sahai, and S. Mitter. Coding into a source: a direct inverse rate-distortion theorem. In *Proceedings of the 44th Annual Allerton Conference on Communication, Control and Computing*, 2003.
- [10] D. Wang, A. Ingber, and Y. Kochman. The dispersion of joint source-channel coding. In *49th Annual Allerton Conference on Communication, Control and Computing*, Monticello, IL, Sep. 2011.
- [11] Te Sun Han. *Information-Spectrum Method in Information Theory*. Springer, 1 edition, November 2002.
- [12] I Csiszár. Joint source-channel error exponent. *Prob. of Cont. and Info. Th.*, 9(5):315–328, 1980.
- [13] I. Csiszár. On the error exponent of source-channel transmission with a distortion threshold. *IEEE Trans. Info. Theory*, IT-28:823–838, Nov. 1982.
- [14] Z. Zhang, E.H. Yang, and V. Wei. The redundancy of source coding with a fidelity criterion - Part one: Known statistics. *IEEE Trans. Info. Theory*, IT-43:71–91, Jan. 1997.