Or Ordentlich Joint work with Uri Erez and Bobak Nazer

Information Theory Workshop Jerusalem, Israel April 27, 2015



$$Y = h_1 X_1 + h_2 X_2 + Z$$

- < ∃ →



$$Y = h_1 X_1 + h_2 X_2 + Z$$

Capacity Region

$$egin{aligned} R_1 &< rac{1}{2} \log(1+h_1^2 ext{SNR}) \ R_2 &< rac{1}{2} \log(1+h_2^2 ext{SNR}) \ R_1 + R_2 &< rac{1}{2} \log(1+\|\mathbf{h}\|^2 ext{SNR}) \end{aligned}$$



< ∃⇒

æ



Feedback Capacity Region (Ozarow 84)

$$egin{aligned} &R_1 < rac{1}{2}\log(1+(1-
ho^2)h_1^2 ext{SNR})\ &R_2 < rac{1}{2}\log(1+(1-
ho^2)h_2^2 ext{SNR})\ &R_1+R_2 < rac{1}{2}\log(1+(\|\mathbf{h}\|^2+2
ho|h_1h_2|) ext{SNR}) \end{aligned}$$



$$Y = h_1 X_1 + h_2 X_2 + Z$$

B K 4 B K



$$Y = h_1 X_1 + h_2 X_2 + Z$$

• We assume full CSI everywhere

- ∢ ≣ ≯



$$Y = h_1 X_1 + h_2 X_2 + Z$$

- We assume full CSI everywhere
- Only lower and upper bounds are known (Nazer & Gastpar 11)

$$\frac{1}{2}\log\left(\frac{1}{2}+\min\{h_1^2,h_2^2\}\mathsf{SNR}\right) \le C_{\mathsf{comp}} \le \frac{1}{2}\log\left(1+\min\{h_1^2,h_2^2\}\mathsf{SNR}\right)$$

• At high SNR the bounds coincide. At low SNR separation is optimal





How Much Does Feedback Help?

 - ∢ ≣ ▶

æ



How Much Does Feedback Help?

• Upper bound remains the same $C_{\text{comp}} \leq \frac{1}{2} \log (1 + \min\{h_1^2, h_2^2\} \text{SNR})$

→ ∃ >



How Much Does Feedback Help?

- Upper bound remains the same $C_{\text{comp}} \leq \frac{1}{2} \log \left(1 + \min\{h_1^2, h_2^2\} \text{SNR}\right)$
- No non-trivial lower bounds are known

→ ∃ >



How Much Does Feedback Help?

- Upper bound remains the same $C_{\text{comp}} \leq \frac{1}{2} \log (1 + \min\{h_1^2, h_2^2\} \text{SNR})$
- No non-trivial lower bounds are known
- In this work we derive a novel lower bound

- 4 ⊒ ▶

Main Result

For any $0<\rho\leq 1$ let

$$\begin{split} \rho_1 &= 1 - (1 - \rho) \left(\frac{h_2}{h_1}\right)^2, \\ R_c &= \frac{1}{2} \log^+ \left(\frac{1}{2} + (1 - \rho) h_2^2 \text{SNR}\right), \\ R' &= \frac{1}{2} \log \left(1 + \frac{(h_1 \sqrt{\rho_1} + h_2 \sqrt{\rho})^2 \text{SNR}}{1 + 2(1 - \rho) h_2^2 \text{SNR}}\right) \end{split}$$

Any computation rate satisfying

$$R < \max_{0 < \rho \leq 1} \min\left(R' + R_c, \frac{1}{2}\log\left(1 + (1 - \rho)h_2^2 \mathsf{SNR}\right)\right)$$

is achievable with feedback.

.

ヘロン 人間 とくほど くほとう







うへつ





Fine lattice Λ_c

프 🕨 🗉 프



Fine lattice Λ_c , coarse lattice $\Lambda \subseteq \Lambda_c$

$$\mathcal{C} = \Lambda_c \cap \mathcal{V}$$

∃ >



Fine lattice Λ_c , coarse lattice Λ , intermediate lattice Λ_s , $\Lambda \subseteq \Lambda_s \subseteq \Lambda_c$

$$\mathcal{C} = \Lambda_c \cap \mathcal{V}$$

伺下 イヨト イヨト



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > C



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > C



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > C



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > C



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > CDecode a list of codewords: $L = {\mathbf{c} \in \mathcal{C} : \mathbf{c} \in [\mathbf{y} + \mathcal{V}_s] \mod \Lambda}$

A 3 b



AWGN channel $\mathbf{y} = \mathbf{x} + \mathbf{z}$, R > CDecode a list of codewords: $L = {\mathbf{c} \in C : \mathbf{c} \in [\mathbf{y} + \mathcal{V}_s] \mod \Lambda}$ $|L| = \log \left(\frac{\operatorname{Vol}(\mathcal{V}_s)}{\operatorname{Vol}(\mathcal{V}_c)}\right)$



Theorem (Song & Devroye 13)

It is possible to decode a list with size $2^{n(R-C)}$ that contains the true codeword w.h.p. using a lattice list decoder

< 注入 < 注入 -

Block Markov coding

ヨト ・ヨト

- Block Markov coding
- In the end of each block user *i* can decode $\mathbf{w}_{\overline{i}}$ using the feedback link

글 제 제 글 제

- Block Markov coding
- In the end of each block user *i* can decode $\mathbf{w}_{\overline{i}}$ using the feedback link
- In each block, each user superimposes encoding of a new message and encoding of the sum of messages from the last block
- The encoding of the sum is transmitted **coherently**

A 3 1 A 3 1

- Block Markov coding
- In the end of each block user *i* can decode $\mathbf{w}_{\overline{i}}$ using the feedback link
- In each block, each user superimposes encoding of a new message and encoding of the sum of messages from the last block
- The encoding of the sum is transmitted **coherently**
- The receiver decodes the coherent part first, and then a list of candidates for the new sum

(B) (B)

- Block Markov coding
- In the end of each block user *i* can decode $\mathbf{w}_{\overline{i}}$ using the feedback link
- In each block, each user superimposes encoding of a new message and encoding of the sum of messages from the last block
- The encoding of the sum is transmitted **coherently**
- The receiver decodes the coherent part first, and then a list of candidates for the new sum

A compute-and-forward variant of Cover-Leung 81

E + 4 E +

For simplicity assume $h_1 = h_2 = 1$

• Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k = 1, \dots, N$ over N+1 blocks

白 マイト イロマー

For simplicity assume $h_1 = h_2 = 1$

- Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k = 1, \dots, N$ over N+1 blocks
- Both users encode their messages using the same lattice code C, such that $\tilde{\mathbf{x}}_i^{(k)} = f\left(\mathbf{w}_i^{(k)}\right) \in C$

御 と く ヨ と く ヨ と ……

For simplicity assume $h_1 = h_2 = 1$

- Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k=1,\ldots,N$ over N+1 blocks
- Both users encode their messages using the same lattice code C, such that $\tilde{\mathbf{x}}_{i}^{(k)} = f\left(\mathbf{w}_{i}^{(k)}\right) \in C$
- First block: $\mathbf{x}_i^{(1)} = \sqrt{1-\rho} \tilde{\mathbf{x}}_i^{(1)}$

通 とう きょう うちょう

For simplicity assume $h_1 = h_2 = 1$

- Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k=1,\ldots,N$ over N+1 blocks
- Both users encode their messages using the same lattice code C, such that $\tilde{\mathbf{x}}_{i}^{(k)} = f\left(\mathbf{w}_{i}^{(k)}\right) \in C$
- First block: $\mathbf{x}_i^{(1)} = \sqrt{1-\rho} \tilde{\mathbf{x}}_i^{(1)}$
- Receiver sees

$$\mathbf{y}^{(1)} = \sqrt{1-\rho} \left(\tilde{\mathbf{x}}_{1}^{(1)} + \tilde{\mathbf{x}}_{2}^{(1)} \right) + \mathbf{z}^{(1)}.$$

For simplicity assume $h_1 = h_2 = 1$

- Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k=1,\ldots,N$ over N+1 blocks
- Both users encode their messages using the same lattice code C, such that $\tilde{\mathbf{x}}_{i}^{(k)} = f\left(\mathbf{w}_{i}^{(k)}\right) \in C$
- First block: $\mathbf{x}_i^{(1)} = \sqrt{1-\rho} \tilde{\mathbf{x}}_i^{(1)}$
- Receiver sees

$$\mathbf{y}^{(1)} = \sqrt{1-\rho} \left(\tilde{\mathbf{x}}_{1}^{(1)} + \tilde{\mathbf{x}}_{2}^{(1)} \right) + \mathbf{z}^{(1)}.$$

• $R < R_{\text{comp}} \triangleq \frac{1}{2} \log \left(\frac{1}{2} + (1 - \rho) \text{SNR} \right)$ is needed for decoding $\mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$

伺下 イヨト イヨト

For simplicity assume $h_1 = h_2 = 1$

- Decoding $\mathbf{w}_1^{(k)} \oplus \mathbf{w}_2^{(k)}, \ k=1,\ldots,N$ over N+1 blocks
- Both users encode their messages using the same lattice code C, such that $\tilde{\mathbf{x}}_i^{(k)} = f\left(\mathbf{w}_i^{(k)}\right) \in C$
- First block: $\mathbf{x}_i^{(1)} = \sqrt{1-\rho} \tilde{\mathbf{x}}_i^{(1)}$
- Receiver sees

$$\mathbf{y}^{(1)} = \sqrt{1-\rho} \left(\tilde{\mathbf{x}}_1^{(1)} + \tilde{\mathbf{x}}_2^{(1)} \right) + \mathbf{z}^{(1)}.$$

- $R < R_{\text{comp}} \triangleq \frac{1}{2} \log \left(\frac{1}{2} + (1 \rho) \text{SNR} \right)$ is needed for decoding $\mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$
- In our case $R > R_{\text{comp}}$ and the receiver can decode a list $L^{(1)}$ of candidates for $\mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$ with size $|L^{(1)}| = 2^{n(R-R_{\text{comp}})}$

• Using the feedback link, user *i* can decode $\mathbf{w}_{i}^{(1)}$ if

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

御 と く ほ と く ほ と

• Using the feedback link, user *i* can decode $\mathbf{w}_{i}^{(1)}$ if

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

• Both users can compute $\mathbf{v}^{(1)} = \mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$

• Using the feedback link, user *i* can decode $\mathbf{w}_{i}^{(1)}$ if

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

- Both users can compute $\mathbf{v}^{(1)} = \mathbf{w}^{(1)}_1 \oplus \mathbf{w}^{(1)}_2$
- Both users apply the same binning function $\mathbf{B} : [2^{nR}] \mapsto [2^{nR'}]$, R' < R, to obtain $\mathbf{B} (\mathbf{v}^{(1)})$

Image: A image: A

• Using the feedback link, user *i* can decode $\mathbf{w}_{i}^{(1)}$ if

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

- Both users can compute $\mathbf{v}^{(1)} = \mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$
- Both users apply the same binning function $\mathbf{B} : [2^{nR}] \mapsto [2^{nR'}]$, R' < R, to obtain $\mathbf{B} (\mathbf{v}^{(1)})$
- Each user encodes $B(v^{(1)})$ to $x^{(1)}_{cohr}$ using the same codebook C' with rate R' and average power SNR

• Using the feedback link, user *i* can decode $\mathbf{w}_{i}^{(1)}$ if

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

- Both users can compute $\mathbf{v}^{(1)} = \mathbf{w}_1^{(1)} \oplus \mathbf{w}_2^{(1)}$
- Both users apply the same binning function $\mathbf{B} : [2^{nR}] \mapsto [2^{nR'}]$, R' < R, to obtain $\mathbf{B} (\mathbf{v}^{(1)})$
- Each user encodes $\mathbf{B}(\mathbf{v}^{(1)})$ to $\mathbf{x}_{cohr}^{(1)}$ using the same codebook \mathcal{C}' with rate R' and average power SNR
- In addition, each user encodes a new message $\mathbf{w}_i^{(2)}$ to the codeword $\tilde{\mathbf{x}}_i^{(2)}$ and transmits

$$\mathbf{x}_{i}^{(2)} = \sqrt{\rho} \mathbf{x}_{cohr}^{(1)} + \sqrt{1-\rho} \tilde{\mathbf{x}}_{i}^{(2)}$$

• • = • • = •

• Channel output is

$$\mathbf{y}^{(2)} = 2\sqrt{\rho}\mathbf{x}_{\mathsf{cohr}}^{(1)} + \sqrt{1-\rho}\left(\tilde{\mathbf{x}}_{1}^{(2)} + \tilde{\mathbf{x}}_{2}^{(2)}\right) + \mathbf{z}^{(2)}$$

回 と く ヨ と く ヨ と

• Channel output is

$$\mathbf{y}^{(2)} = 2\sqrt{\rho}\mathbf{x}_{cohr}^{(1)} + \sqrt{1-\rho}\left(\tilde{\mathbf{x}}_{1}^{(2)} + \tilde{\mathbf{x}}_{2}^{(2)}\right) + \mathbf{z}^{(2)}$$

 $\bullet~\mbox{Can}~\mbox{decode}~ {\bf x}^{(1)}_{\mbox{cohr}}~\mbox{if}$

$$R' \leq rac{1}{2} \log \left(1 + rac{4
ho \mathsf{SNR}}{1 + 2(1 -
ho) \mathsf{SNR}}
ight)$$

→ E → < E →</p>

• Channel output is

$$\mathbf{y}^{(2)} = 2\sqrt{\rho}\mathbf{x}_{\mathsf{cohr}}^{(1)} + \sqrt{1-\rho}\left(\tilde{\mathbf{x}}_{1}^{(2)} + \tilde{\mathbf{x}}_{2}^{(2)}\right) + \mathbf{z}^{(2)}$$

• Can decode $\mathbf{x}_{cohr}^{(1)}$ if

$${\mathcal R}' \leq rac{1}{2} \log \left(1 + rac{4
ho {\mathsf{SNR}}}{1+2(1-
ho){\mathsf{SNR}}}
ight)$$

• The decoder looks for a unique $\mathbf{w} \in \mathbb{F}_p^k$ in $L^{(1)} \cap \mathbf{B}^{-1}\left(\mathbf{v}^{(1)}\right)$

글 제 제 글 제

• Channel output is

$$\mathbf{y}^{(2)} = 2\sqrt{\rho}\mathbf{x}_{cohr}^{(1)} + \sqrt{1-\rho}\left(\tilde{\mathbf{x}}_{1}^{(2)} + \tilde{\mathbf{x}}_{2}^{(2)}\right) + \mathbf{z}^{(2)}$$

• Can decode $\mathbf{x}_{cohr}^{(1)}$ if

$${{\mathcal{R}}'} \le rac{1}{2}\log \left(1 + rac{4
ho {\mathsf{SNR}}}{1 + 2(1 -
ho) {\mathsf{SNR}}}
ight)$$

The decoder looks for a unique w ∈ 𝔽^k_p in L⁽¹⁾ ∩ B⁻¹ (v⁽¹⁾)
If R' > R − R_{comp} such a w ∈ 𝔽^k_p will be found with probability 1

• Channel output is

$$\mathbf{y}^{(2)} = 2\sqrt{\rho}\mathbf{x}_{cohr}^{(1)} + \sqrt{1-\rho}\left(\tilde{\mathbf{x}}_{1}^{(2)} + \tilde{\mathbf{x}}_{2}^{(2)}\right) + \mathbf{z}^{(2)}$$

• Can decode $\mathbf{x}_{cohr}^{(1)}$ if

$${\mathcal R}' \leq rac{1}{2} \log \left(1 + rac{4
ho {\mathsf{SNR}}}{1+2(1-
ho){\mathsf{SNR}}}
ight)$$

- The decoder looks for a unique $\mathbf{w} \in \mathbb{F}_p^k$ in $L^{(1)} \cap \mathbf{B}^{-1}\left(\mathbf{v}^{(1)}\right)$
- If $R' > R R_{comp}$ such a $\mathbf{w} \in \mathbb{F}_p^k$ will be found with probability 1
- Next, the decoder subtracts $\mathbf{x}_{cohr}^{(1)}$ from $\mathbf{y}^{(2)}$ and decodes a list $L^{(2)}$ of candidates for $\mathbf{v}^{(2)} = \mathbf{w}_1^{(2)} \oplus \mathbf{w}_2^{(2)}$

• Correct decoding through feedback link

$$R < rac{1}{2}\log(1+(1-
ho){\mathsf{SNR}})$$

Image: A image: A

• Correct decoding through feedback link

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

• Correct decoding of \mathbf{x}_{cohr}

$$R' \leq rac{1}{2}\log\left(1 + rac{4
ho \mathsf{SNR}}{1 + 2(1 -
ho)\mathsf{SNR}}
ight)$$

• 3 3

• Correct decoding through feedback link

$$R < rac{1}{2}\log(1+(1-
ho)\mathsf{SNR})$$

• Correct decoding of \boldsymbol{x}_{cohr}

$$R' \leq rac{1}{2} \log \left(1 + rac{4
ho \mathsf{SNR}}{1 + 2(1 -
ho) \mathsf{SNR}}
ight)$$

• Unique element in intersection of list and bin

$$R' > R - \frac{1}{2}\log\left(\frac{1}{2} + (1-
ho)\mathsf{SNR}\right)$$

Achievable Rate

$$\begin{split} R &< \min\left\{\frac{1}{2}\log(1+(1-\rho)\mathsf{SNR}), \\ & \frac{1}{2}\log\left(1+\frac{4\rho\mathsf{SNR}}{1+2(1-\rho)\mathsf{SNR}}\right) + \frac{1}{2}\log\left(\frac{1}{2}+(1-\rho)\mathsf{SNR}\right)\right\} \end{split}$$

通 と く ヨ と く ヨ と

Achievable Rate

$$egin{aligned} R < \min\left\{rac{1}{2}\log(1+(1-
ho)\mathsf{SNR}), & \ & rac{1}{2}\log\left(rac{1}{2}+(1+
ho)\mathsf{SNR}
ight)
ight\} \end{aligned}$$

回 と く ヨ と く ヨ と

Achievable Rate

$$egin{aligned} R < \min\left\{rac{1}{2}\log(1+(1-
ho)\mathsf{SNR}), \ & rac{1}{2}\log\left(rac{1}{2}+(1+
ho)\mathsf{SNR}
ight)
ight\} \end{aligned}$$

Setting $\rho = \frac{1}{4\text{SNR}}$ we get

$$R < \frac{1}{2} \log \left(\frac{3}{4} + \mathsf{SNR} \right)$$

▶ < 프 ▶ < 프 ▶</p>

- We studied the problem of computing a linear function from the output of a Gaussian MAC with feedback
- We derived a new coding scheme for this scenario
- For a symmetric setting our scheme achieves $R = \frac{1}{2} \log \left(\frac{3}{4} + SNR \right)$
- The scheme can be extended to noisy feedback and more than 2 users
- Our scheme works in blocks. Can we find a scalar, à la Schalkwijk-Kailath 66 scheme?

通 とう きょう うちょう